

In the Name of GOD

Digital Modulations

In this session we want to talk about digital Modulation techniques and we want to analyse the main parameters of signal space & equi. vector space for these Modulation.

1) Amplitude Modulation (ASK)

Amplitude Shift Keying or Pass-band PAM

(Pulse Amplitude Modulation)

As you know in ASK Modulation, our digital information modulate the amplitude of carrier waveform. Means that based on the information symbols in the input of modulator

a related amplitude is selected and then the carrier waveform is transmitted from transmitter to the receiver using this amplitude.

So, the signals we use for transmission using m -ary ASK modulation $\{S_m(t)\}_{m=1}^m$ are as follows;

$$S_m(t) = \underbrace{A_m}_{\substack{\text{Amplitude} \\ \text{of carrier signal} \\ \text{that carry our} \\ \text{information}}} \underbrace{g(t)}_{\downarrow} \underbrace{C \cdot 2\pi f_c t}_{\substack{\text{Carrier waveform} \\ \text{Carrier frequency}}} \quad 1 \leq m \leq M, \quad 0 \leq t \leq \underbrace{T}_{\substack{\text{Symbol} \\ \text{duration}}}$$

$g(t)$ is an specified signal used to pulse-shaping or spectrum-shaping of trasmitted signal proper for Comm. channel characteristics

As you can see from the relation of $\{s_m(t)\}_{m=1}^M$,
all signals in this set are proportion to $g(t) e^{j2\pi f_c t}$.

Mean all signals in this set, are ^{presented a} the linear combination of
 $g(t) e^{j2\pi f_c t}$. So, based on what we learn from Gram-Schmidt
algorithm, we can select $g(t) e^{j2\pi f_c t}$ and normalize it

and take it as the basis function of the signal
Set spanned by $\{s_m(t)\}_{m=1}^M$. So, we have just one
basis as follows,

$$f_1(t) = \frac{g(t) \cos 2\pi f_c t}{\|g(t) \cos 2\pi f_c t\|}$$

and the dimension of signal space is equal to the number of basis functions and so is equal to 1. ($N=1$)
we should calculate $\|g(t) e^{j2\pi f_c t}\|$ to completely define $f_s(t)$ as the signal space basis function.

$$\|g(t) e^{j2\pi f_c t}\|^2 = \int_0^T \left(g(t) e^{j2\pi f_c t}\right)^2 dt$$

$$\Rightarrow \int_0^T g^2(t) \underbrace{\cos^2 2\pi f_c t}_{\frac{1}{2}(1 + \cos 4\pi f_c t)} dt = \underbrace{\frac{1}{2} \int_0^T g^2(t) dt}_{\substack{\checkmark \\ \text{Energy of } g(t) = E_g}} + \underbrace{\frac{1}{2} \int_0^T g^2(t) \cos 4\pi f_c t dt}_0$$

narrow-band signal

$$\Rightarrow \|g(t) \cos 2\pi f_c t\|^2 = \frac{1}{2} E_g \quad \Rightarrow \|g(t) \cos 2\pi f_c t\| = \sqrt{\frac{E_g}{2}}$$

$$\Rightarrow f_1(t) = \frac{g(t) \cos 2\pi f_c t}{\|g(t) \cos 2\pi f_c t\|} = \underbrace{\sqrt{\frac{2}{E_g}} g(t) \cos 2\pi f_c t}_{\text{Signal space basis function}}$$

So, we can represent all signals in the signal space such^{as}

$\{s_m(t)\}_{m=1}^M$ as the linear combination of $f_1(t)$ to get

the equivalent vector space.

$$S_m(t) = A_m g(t) \cos 2\pi f_c t$$

$$1 \leq m \leq M, \quad 0 \leq t \leq T \quad (1)$$

$$f_1(t) = \sqrt{\frac{2}{E_g}} g(t) \cos 2\pi f_c t$$

(2)

(1), (2) $\Rightarrow$

$$S_m(t) = \underbrace{\sqrt{\frac{E_g}{2}} A_m}_{S_m} f_1(t)$$

$$S_m = S_m = \sqrt{\frac{E_g}{2}} A_m$$

Signal Constellation for M-ASK

Now, we want to analyse the signal space & vector space for M-ASK modulation technique and calculate the main parameters of this signalling set.

First parameter is signal energies $\{S_m(t)\}_{m=1}^M$

$$S_m(t) = A_m g(t) \cos 2\pi f_c t = S_m f_1(t)$$

in which $S_m = \sqrt{\frac{E_g}{2}} A_m$ and $f_1(t) = \sqrt{\frac{2}{E_g}} g(t) \cos 2\pi f_c t$

As we know the energy of $s_m(t)$ is defined as follows

$$\underbrace{E_m}_{\substack{\downarrow \\ \text{Energy of} \\ s_m(t)}} = \int_0^T s_m^2(t) dt = \int_0^T (A_m g(t) \cos 2\pi f_c t)^2 dt = \|\underline{s}_m\|^2 = \frac{\epsilon_g}{2} A_m^2$$

$s_m(t) = s_m f_c(t)$

$s_m = \sqrt{\frac{\epsilon_g}{2}} A_m$

$\uparrow$
Vector
Space

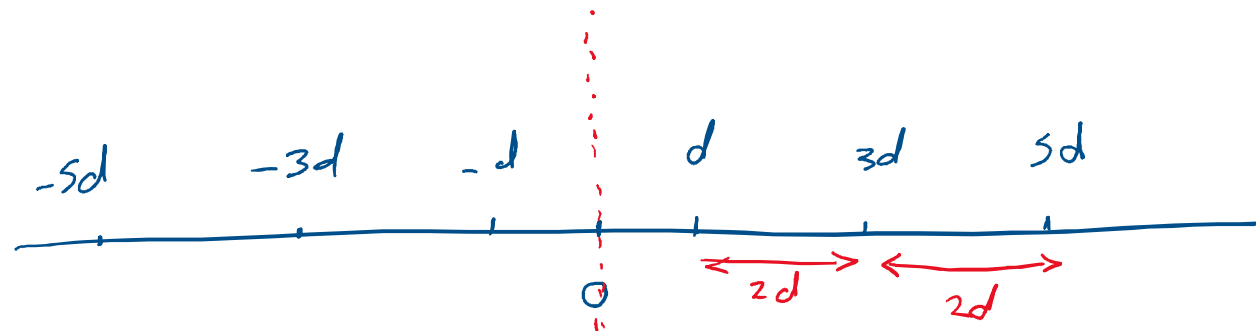
In digital Modulation techniques, it is desirable to have a symmetric signal Constellation. Because this symmetry can help us to design the modulator & demodulator in a simple manner.

So, in M-ASK modulation $\{A_m\}_{m=1}^M$ are selected such as the signal space symmetry is satisfied. Therefore, $\{A_m\}_{m=1}^M$ are usually selected as below;

$$A_m = (2m-1-d)d, \quad 1 \leq m \leq M$$

$$\Rightarrow A_m \in \{\pm d, \pm 3d, \pm 5d, \pm 7d, \dots, \pm (M-1)d\}$$

modulation
levels



The distance between any two adjacent modulation level is equal to $2d$.

Symmetric around zero

$$\Rightarrow E_m = \frac{E_g}{2} A_m^2 = \frac{E_g}{2} (2m-1-m)^2 d^2$$

second parameter is the average energy of signals $\{s_n(t)\}_{n=1}^M$

The average energy of signals $\{s_n(t)\}_{n=1}^M$ in m-ASK mod. is as follows

$$\underbrace{E_{ave}}_{\downarrow} = \frac{1}{M} \sum_{m=1}^M E_m = \frac{1}{M} \sum_{m=1}^M \frac{E_g}{2} \overset{2}{A_m^2} = \frac{E_g}{2M} \sum_{m=1}^M A_m^2$$



$$E_m = \frac{E_g}{2} A_m^2$$

The average
energy of
signals

$$A_m = (2m-1-\pi)d \implies$$

$$E_{ave} = \frac{E_g}{2M} \sum_{m=1}^M (2m-1-\pi)^2 d^2 = \frac{E_g d^2}{2M} \sum_{m=1}^M (2m-1-\pi)^2$$

As we know $(2m-1-M) \Big|_{m=1}^M = \{ \pm d, \pm 3d, \pm 5d, \dots, \pm (M-1)d \}$

$$\Rightarrow E_{ave} = \frac{E_g d^2}{2M} \underbrace{\sum_{m=1}^M (2m-1-M)^2}_{\text{Summation of square of odd numbers between 1 to } M-1 \text{ with } + \text{ and } - \text{ signs.}} = \frac{E_g d^2}{2M} \times \cancel{2} \times \frac{1}{6} \cancel{M(M-1)(M+1)}$$

$2 \times \frac{1}{6} M(M-1)(M+1) \leftarrow$

$$\Rightarrow E_{ave} = \frac{E_g d^2 (M^2 - 1)}{6}$$

: The average Energy of
Signals in M-ASK Modulation

Third Parameter is the average Energy Per bit

We know that in M-ary modulation each symbol contain $k = \log_2 M$ bits. So, if we want to compare different M-ary modulations for

different values of M , it is not fair to compare them
based on ^{ave.} symbol energy.

It is fair when we compare them

based on ave. energy per bit. So, we should calculate the

ave. energy per bit for M -ary mod. techniques

$$E_{b_{ave}} = \frac{E_{ave}}{k} = \frac{E_{ave}}{\log_2^M} = \frac{E_g d^2 (M^2 - 1)}{6 \log_2^M}$$

average energy
per bit for
 M -ASK

4th Parameter is Euclidean distance between any two signals

The Euclidean distance between any two signal $s_m(t)$ and $s_n(t)$ in the M-ASK signal space is calculated as

below;

$$d_{mn}^{(e)} = \| s_m(t) - s_n(t) \| = \overset{\text{vector space}}{\downarrow} \| \underline{s}_m - \underline{s}_n \|$$

$$d_{mn}^{(e)} = \| S_m(t) - S_n(t) \| = \left[\int_0^T (S_m(t) - S_n(t))^2 dt \right]^{1/2}$$

$$= \| \underline{S}_m - \underline{S}_n \| = | S_m - S_n | = \left| \sqrt{\frac{\epsilon_g}{2}} A_m - \sqrt{\frac{\epsilon_g}{2}} A_n \right|$$

↑

vector
space

$$S_m(t) = A_m g(t) \cos 2\pi f_c t = S_m f_1(t) \quad , \quad S_m = \sqrt{\frac{\epsilon_g}{2}} A_m$$

$$A_m = (2m-1-M)d$$

$$S_n(t) = A_n g(t) \cos 2\pi f_c t = S_n f_1(t) \quad , \quad S_n = \sqrt{\frac{\epsilon_g}{2}} A_n$$

$$A_n = (2n-1-M)d$$

$$\Rightarrow d_{mn}^{(e)} = \sqrt{\frac{\epsilon_g}{2}} |A_m - A_n| = \sqrt{\frac{\epsilon_g}{2}} |2(n-m)d| = \sqrt{2\epsilon_g} |n-m|d$$

$\uparrow$
 $(2m-1-m)d$
 $\uparrow$
 $(2n-1-m)d$

$$\Rightarrow d_{mn}^{(e)} = \sqrt{2\epsilon_g} |n-m|d \quad : \text{Euclidean distance between } S_m(t) \text{ and } S_n(t)$$

5th Parameter is minimum Euclidean distance between signals

As you can conclude from signal rep. in vector space, the minimum Euclidean distance between signals in M-ASK modulation is where the signals s_{m+1} & s_{n+1} are adjacent signals. Means $|m-n| = 1$

$$\Rightarrow d_{\min} = \sqrt{2E_g} d = d \sqrt{2E_g}$$

$\downarrow$

for M-ASK
Signalling

Minimum Euclidean distance
between Signals

We will see that $d_{\min}$ is one of the main factors that affect the modulation performance.

6th parameter is the relation between $d_{\min}$ and E_{bave}

To fairly comparison of different signalling with different values of M , we should do the comparison based

E_{bave} (average energy per bit). on the other hand, $d_{\min}$ is

one of the main factors that affect the performance of signalling.

So, relation between $d_{\min}$ and $E_{b_{av}}$ is important for us to know about each signalling. For M-ASK signalling we can obtain this relation as follows.

$$d_{\min} = d \sqrt{2E_g} \quad (1) \quad \text{and}$$

$$E_{b_{av}} = \frac{E_g d^2 (M^2 - 1)}{6 \log_2 M} \quad (2)$$

Note: In the original image, a red box highlights $E_g d^2$ in equation (2), and a red arrow points from this box to $(d_{\min}^2)/2$ written above it.

I want to find the relation of $d_{\min}$ based on $E_{b_{av}}$

①, ②
 $\Rightarrow$

$$\epsilon_{\text{bave}} = \frac{d_{\min}^2 (M^2 - 1)}{2 \times 6 \lg_2^M}$$

$\Rightarrow$

$$d_{\min} = \sqrt{\frac{12 \lg_2^M}{M^2 - 1}} \epsilon_{\text{bave}}$$

In M-ASK modulation it is important to note to these comments.

As you know in some Comm. systems, we use PAM signalling in the baseband form, means

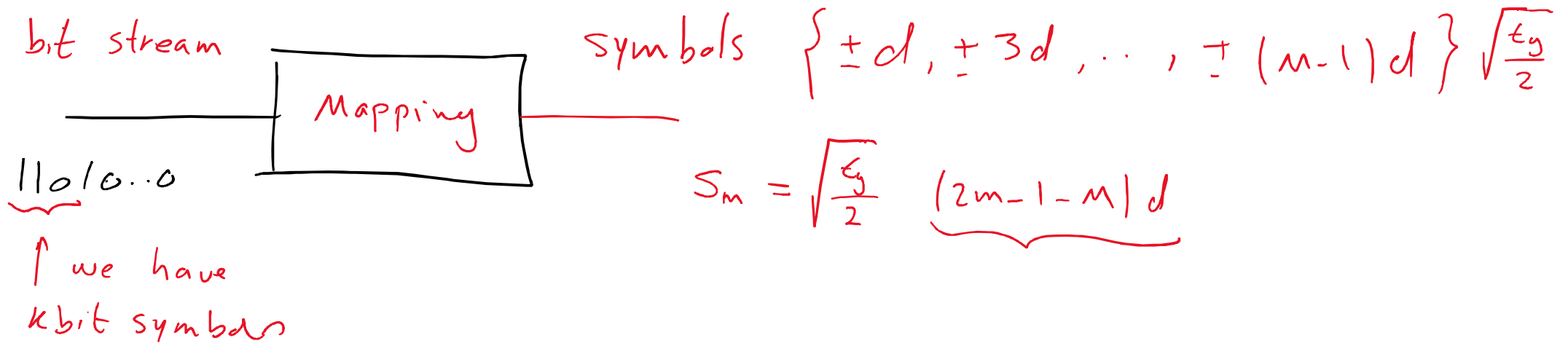
$$s_m(t) = A_m \underbrace{p(t)}_{\text{Pulse-shaping signal}} \quad 1 \leq m \leq M \quad \text{and} \quad 0 \leq t \leq T$$

means that the information is transmitted in the baseband form without any carrier waveform.

Obtain the parameters of baseband as an Exercise.

(Signal Energies, E_{av} , E_{bav} , $d_{min}^{(e)}$, d_{min} , d_{min} and ϵ_{bave} relative)

The other comment is on the mapping assignment of bits to symbols at the M-ASK modulator input.



In M-ASK modulation technique, we assign each k -bits of input bit stream to M-ASK symbols in two different way. The first way is called Natural Coding and the second way is called Gray Coding.

In natural Coding we assign the modulation levels to each k -bits of input stream, in descending (or ascending) order. But in Gray Coding we do assignment such that the bit error probability in detection will be minimized.